

1080 
$$\sum_{q \in Q(m)} \pi_1(q) \leq R_1(m), \gamma(q) \geq 0 \quad ]$$

1081 is solved using the approximation in ~~EQ. 11~~, [

1082 
$$\gamma(q) = \frac{P_{21}(q, q) \pi_1(q)}{i_2(q)} \quad ]$$

1083 and the network further comprises at least one high-level network controller that controls  
 1084 the power constraints  ~~$R_1(q)$~~  [ $R_1(m)$ ], and drives the network towards a max-min  
 1085 solution.

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1088 73. (currently amended) A method as in claim 60[61], wherein each node:

1089 is given an initial  $\gamma_0$ ;

1090 generates the model expressed in ~~EQ. 20, EQ. 21, and EQ. 22~~ [

1091 
$$L(\gamma, g, \beta) = g^T \gamma, \sum_{q \in Q(m)} \log(1 + \gamma(q)) \geq \beta(m)$$

1092 
$$g = \nabla_{\gamma} f(\gamma_0) \quad ];$$

1093 updates the new  $\gamma_{\alpha}$  from ~~EQ. 23 and EQ. 24~~

1094 
$$[ \gamma_* = \arg \min_{\gamma} L(\gamma, g, \beta), \gamma_{\alpha} = \gamma_0 + \alpha(\gamma_* - \gamma_0) \quad ];$$

1095 determines a target SINR to adapt to; and,

1096 updates the transmit power for each link  $q$  according to ~~EQ. 25 and EQ. 26~~ [

1097 
$$\pi_2(q) = \gamma_{\alpha} i_1(q) / |h(q)|^2$$

1098 
$$\pi_1(q) = \gamma_{\alpha} i_2(q) / |h(q)|^2 \quad ].$$

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1100 74. (currently amended) A method as in claim 60[61], for each node wherein the

1101 transmit power relationship of ~~EQ. 25 and EQ. 26~~ [

$$\pi_2(q) = \gamma_\alpha i_1(q) / |h(q)|^2$$

$$\pi_1(q) = \gamma_\alpha i_2(q) / |h(q)|^2 \quad 1$$

1104 is not known, that:

1105 uses a suitably long block of  $N$  samples is used to establish the relationship, where  
 1106  $N$  is either 4 times the number of antennae or 128, whichever is larger;  
 1107 uses the result to update the receive weights at each end of the link;  
 1108 optimizes the local model as in ~~EQ. 23 and EQ. 24~~ [

$$\gamma_* = \arg \min_{\gamma} L(\gamma, \mathbf{g}, \beta)$$

$$\gamma_\alpha = \gamma_0 + \alpha(\gamma_* - \gamma_0) \quad 1;$$

1111 and then applies [

$$\pi_2(q) = \gamma_\alpha i_1(q) / |h(q)|^2$$

$$\pi_1(q) = \gamma_\alpha i_2(q) / |h(q)|^2 \quad 1 \quad \text{EQ. 25 and EQ. 26.}$$

1114  
 1115 75. (currently amended) A method as in claim 60[61] that, for an aggregate proper  
 1116 subset  $m$ :

1117 for each node within the set  $m$ , inherits the network objective function model  
 1118 given in ~~EQ. 28, EQ. 29, and EQ. 30~~ [

$$L_m(\gamma, \mathbf{g}, \beta) = \sum_{q \in Q(m)} \mathbf{g}_q \gamma(q)$$

$$\sum_{q \in Q(m)} \log(1 + \gamma(q)) \geq \beta(m)$$

$$g(q) = i_1(q)i_2(q) / |h(q)|^2 \quad 1;$$

eliminates the [a] step of matrix channel estimation, transmitting instead from that node as a single real number for each link to the other end of said link an estimate of the post beamforming interference power; and , receives back for each link a single real number being the transmit power.

76. (original) A method as in claim 75 [ 74 ], that for each pair of nodes assigns to the one presently possessing the most processing capability the power management computations.

77. (currently amended) A method as in claim 74[75] that estimates the transfer gains and the post beamforming interference power using simple least squares estimation techniques.

78. (currently amended) A method as in claim 74[75] that, for estimating the transfer gains and post beamforming interference power:

instead solves for the transfer gain  $h$  using ~~EQ. 31~~

$$[ y(n) = hgs(n) + \varepsilon(n) ];$$

uses a block of  $N$  samples of data to estimate  $h$  using ~~EQ. 32~~ [

$$h = \frac{\sum_{n=1}^N s^*(n)y(n)}{\sum_{n=1}^N |s(n)|^2 g} ]$$

obtains an estimation of residual interference power  $\hat{R}_e [ R_e ]$  using ~~EQ. 33~~ [

$$R_e = \left\langle \left| \varepsilon(n) \right|^2 \right\rangle$$
$$= \frac{1}{N} \sum_{n=1}^N \left( \left| y(n) \right|^2 - \left| ghs(n) \right|^2 \right)$$

and,

obtains knowledge of the transmitted data symbols  $S(n)$  from using remodulated symbols at the output of the codec.

79. (currently amended) A method as in claim 77 [78] wherein, instead of obtaining knowledge of the transmitted data symbols  $S(n)$  from using remodulated symbols at the output of the codec, the node uses the output of a property restoral algorithm used in a blind beamforming algorithm.

80. (currently amended) A method as in claim 77 [78] wherein, instead of obtaining knowledge of the transmitted data symbols  $S(n)$  from using remodulated symbols at the output of the codec, the node uses a training sequence explicitly transmitted to train beamforming weights and asset the power management algorithms.

81. (currently amended) A method as in claim 77 [78] wherein, instead of obtaining knowledge of the transmitted data symbols  $S(n)$  from using remodulated symbols at the output of the codec, the node uses any combination of:

the output of a property restoral algorithm used in a blind beamforming algorithm;  
a training sequence explicitly transmitted to train beamforming weights and asset the power management algorithms;

or,

1171 other means known to the art.

1172

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1174 82. (currently amended) A method as in claim 60[61], wherein each node  
1175 incorporates a link level optimizer and a decision algorithm, ~~as illustrated in Figure~~  
1176 ~~32A and 32B.~~

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1178 83. (currently amended) A method as in claim 81[82], wherein the decision  
1179 algorithm is a Lagrange multiplier technique.

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1181

1182 84. (currently amended) A method as in claim 60[61], wherein the solution to ~~EQ. 3~~  
1183 
$$\left[ \min_{\pi_1(q)} \sum \pi_1(q) = 1^T \pi_1 \right]$$
 is implemented by a penalty function technique.

1184

1185

1186 85. (currently amended) A method as in claim 83[84], wherein the penalty function  
1187 technique:

1188 takes the derivative of  $\gamma(q)$  [  $\gamma(q)$  ] with respect to  $\pi_1$ ;

1189 and,

1190 uses the Kronecker-Delta function and the weighted background noise.

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1193 86. (currently amended) A method as in claim 83[84], wherein the penalty function  
1194 technique neglects the noise term.

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1197 87. (currently amended) A method as in claim 83[84], wherein the penalty function  
1198 technique normalizes the noise term to one.

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1200  
1201 88. (currently amended) A method as in claim 60[61], wherein the approximation  
1202 uses the receive weights.  
1203

1204  
1205 89. (currently amended) A method as in claim 60[61], wherein adaptation to the  
1206 target objective is performed in a series of measured and quantized descent and ascent  
1207 steps.  
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1209 90. (currently amended) A method as in claim 60[61], wherein the adaptation to the  
1210 target objective is performed in response to information stating the vector of change.  
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1212  
1213 91. (currently amended) A method as in claim 60[61], which uses the log linear  
1214 mode in ~~EQ. 34~~ [

$$\beta_q \approx \log \left( \frac{a \pi_1(q) + a_0}{b \pi_1(q) + b_0} \right) = \hat{\beta}_q(\pi_1(q)) ]$$

1215  
1216 and the inequality characterization in ~~EQ. 35~~ [  $\hat{\beta}_q(\pi_1(q)) \geq \beta$  ] to solve the  
1217 approximation problem with a simple low dimensional linear program.  
1218

1219  
1220 92. (currently amended) A method as in claim 60[61], develops the local mode by  
1221 matching function values and gradients between the current model and the actual  
1222 function.  
1223

1224  
1225 93. (currently amended) A method as in claim 60[61], which develops the model as  
1226 a solution to the least squares fit, evaluated over several points.  
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